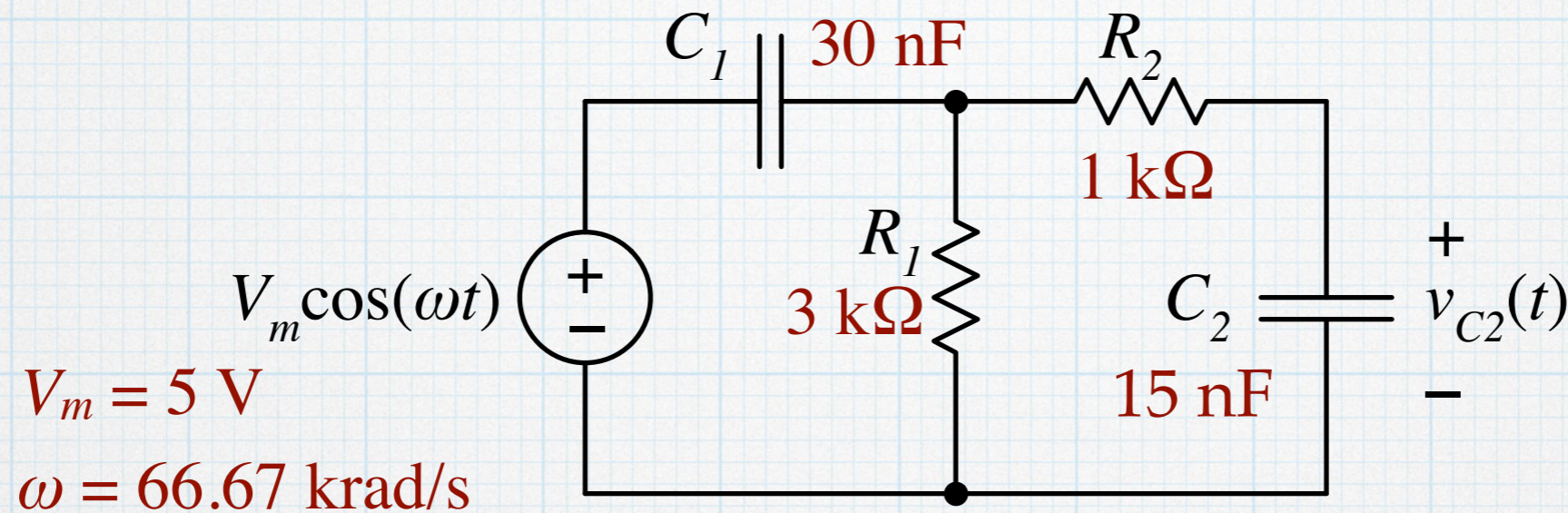
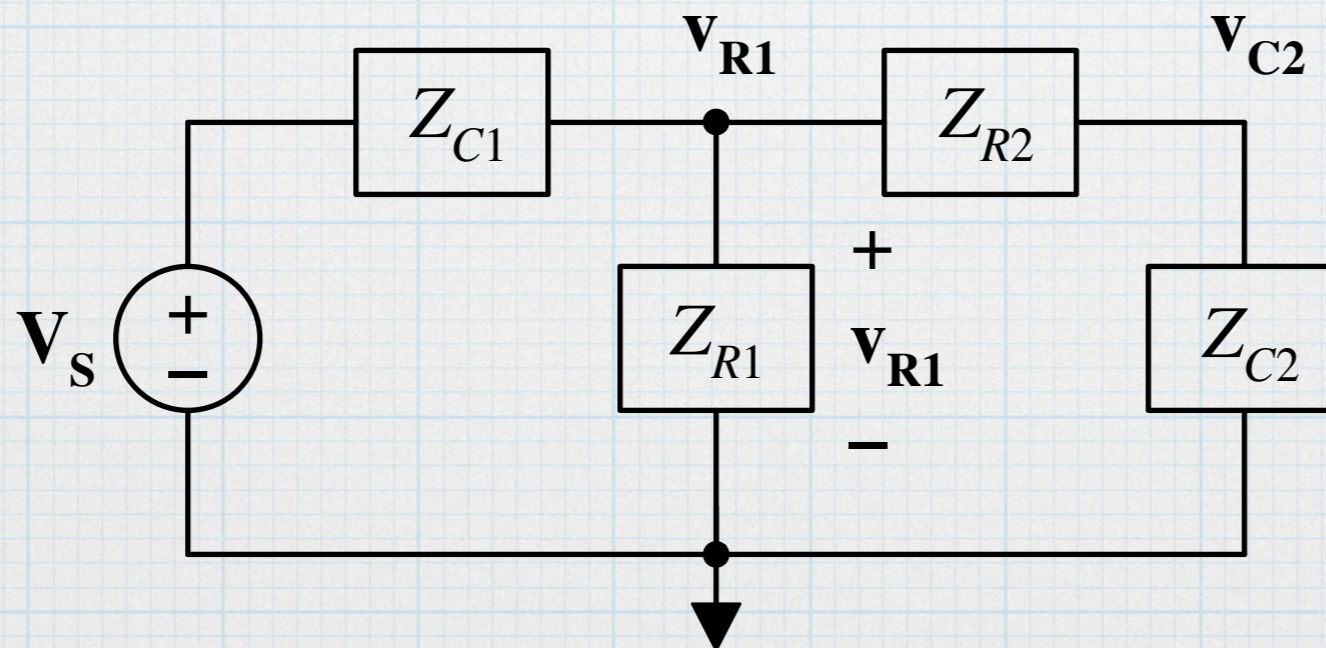


## Example 7 - node voltage.

Same circuit as Example 4, but now solve it using node voltage.



Transform the circuit to its complex equivalent. The node voltages are  $\mathbf{v}_{R1}$  and  $\mathbf{v}_{C2}$ .



Write the two node-voltage equations.

$$\frac{V_s - v_{R1}}{Z_{C1}} = \frac{v_{R1} - v_{C2}}{Z_{R2}} + \frac{v_{R1}}{Z_{R1}}$$

$$\frac{v_{R1} - v_{C2}}{Z_{R2}} = \frac{v_{C2}}{Z_{C2}}$$

Re-arrange a bit;

$$\left[ 1 + \frac{Z_{C1}}{Z_{R2}} + \frac{Z_{C1}}{Z_{R1}} \right] v_{R1} - \frac{Z_{C1}}{Z_{R2}} v_{C2} = V_s$$

$$v_{R1} = \left[ 1 + \frac{Z_{R2}}{Z_{C2}} \right] v_{C2}$$

Substitute:

$$\left[ 1 + \frac{Z_{C1}}{Z_{R2}} + \frac{Z_{C1}}{Z_{R1}} \right] \left[ 1 + \frac{Z_{R2}}{Z_{C2}} \right] v_{C2} - \frac{Z_{C1}}{Z_{R2}} v_{C2} = V_s$$

Substitute in for the impedances

$$\left[ 1 + \frac{1}{j\omega R_2 C_1} + \frac{1}{j\omega R_1 C_1} \right] [1 + j\omega R_2 C_2] v_{C2} - \frac{1}{j\omega R_2 C_1} v_{C2} = V_s$$

Multiply all the terms and then gather real and imaginary parts. (Note that two of the terms cancel.) Grind it out.

$$\left[ \left( 1 + \frac{C_2}{C_1} + \frac{R_2 C_2}{R_1 C_1} \right) + j \left( \omega R_2 C_2 - \frac{1}{\omega R_1 C_1} \right) \right] v_{C2} = V_s$$

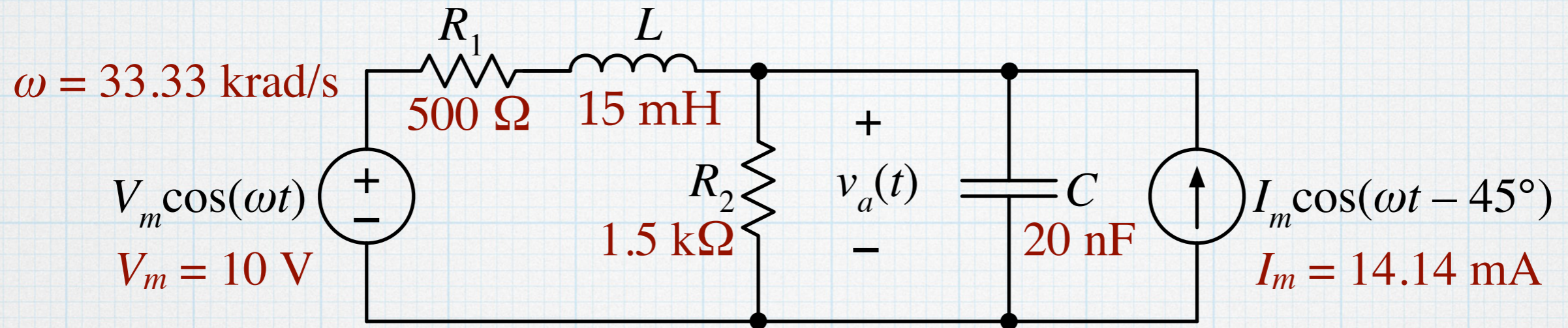
$$[1.667 + j0.8333] v_{C2} = V_s$$

$$\left[ (1.863) e^{26.6^\circ} \right] v_{C2} = 5 \text{ V}$$

$$v_{C2} = (2.68 \text{ V}) e^{-j26.6^\circ} \quad \text{Same as example 4.}$$

## Example 8 - node voltage.

Find the complex value of the capacitor voltage in the “two-source, two-impedance” circuit (Example 6) using the node voltage method.



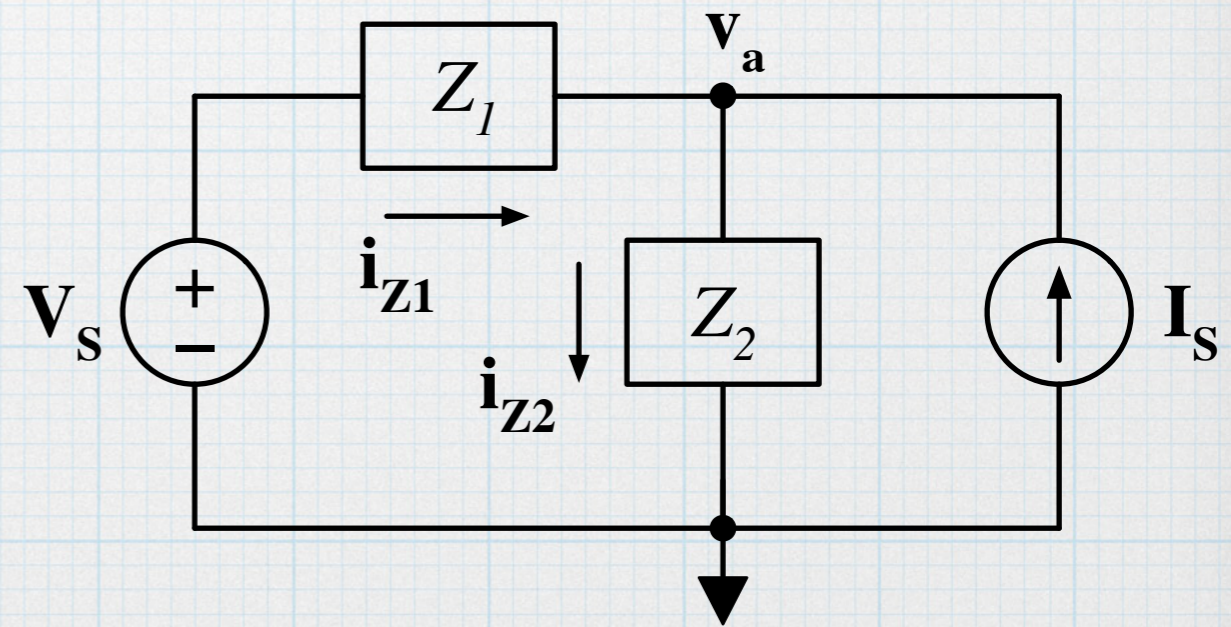
Transform to the complex form of the circuit. The circuit and the values are identical to Example 6.

$$\mathbf{V}_S = V_m e^{j0^\circ} = 10 \text{ V}$$

$$\mathbf{I}_S = I_m e^{-j45^\circ} = (14.14 \text{ mA}) e^{-j45^\circ}$$

$$\mathbf{Z}_1 = (707.1 \Omega) e^{j45^\circ}$$

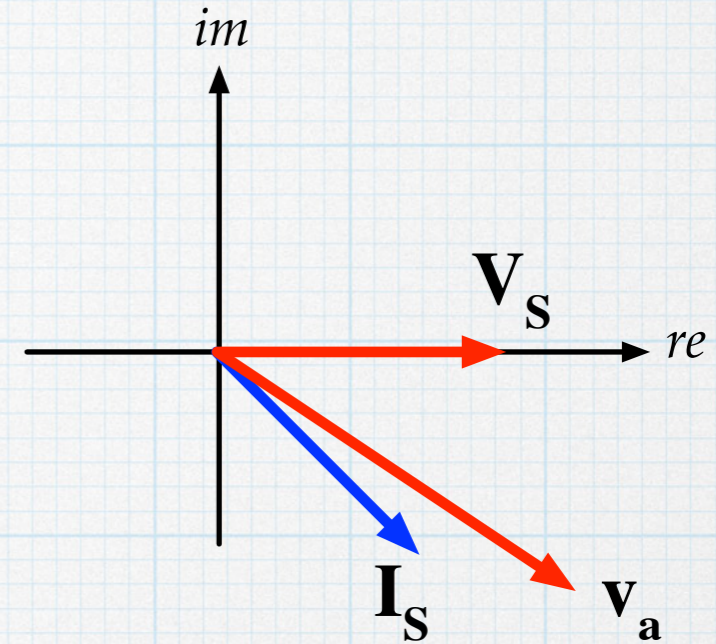
$$\mathbf{Z}_2 = (1060.7 \Omega) e^{-j45^\circ}$$



$$i_{z1} + I_S = i_{z2}$$

$$\frac{V_S - v_a}{Z_1} + I_S = \frac{v_a}{Z_2}$$

$$v_a = \frac{V_S + Z_1 I_S}{1 + \frac{Z_1}{Z_2}}$$



$$= \frac{10 \text{ V} + [(707.1 \Omega) e^{j45^\circ}] [(14.14 \text{ mA}) e^{-j45^\circ}]}{1 + \frac{(707.1 \Omega) e^{j45^\circ}}{(1061 \Omega) e^{-j45^\circ}}}$$

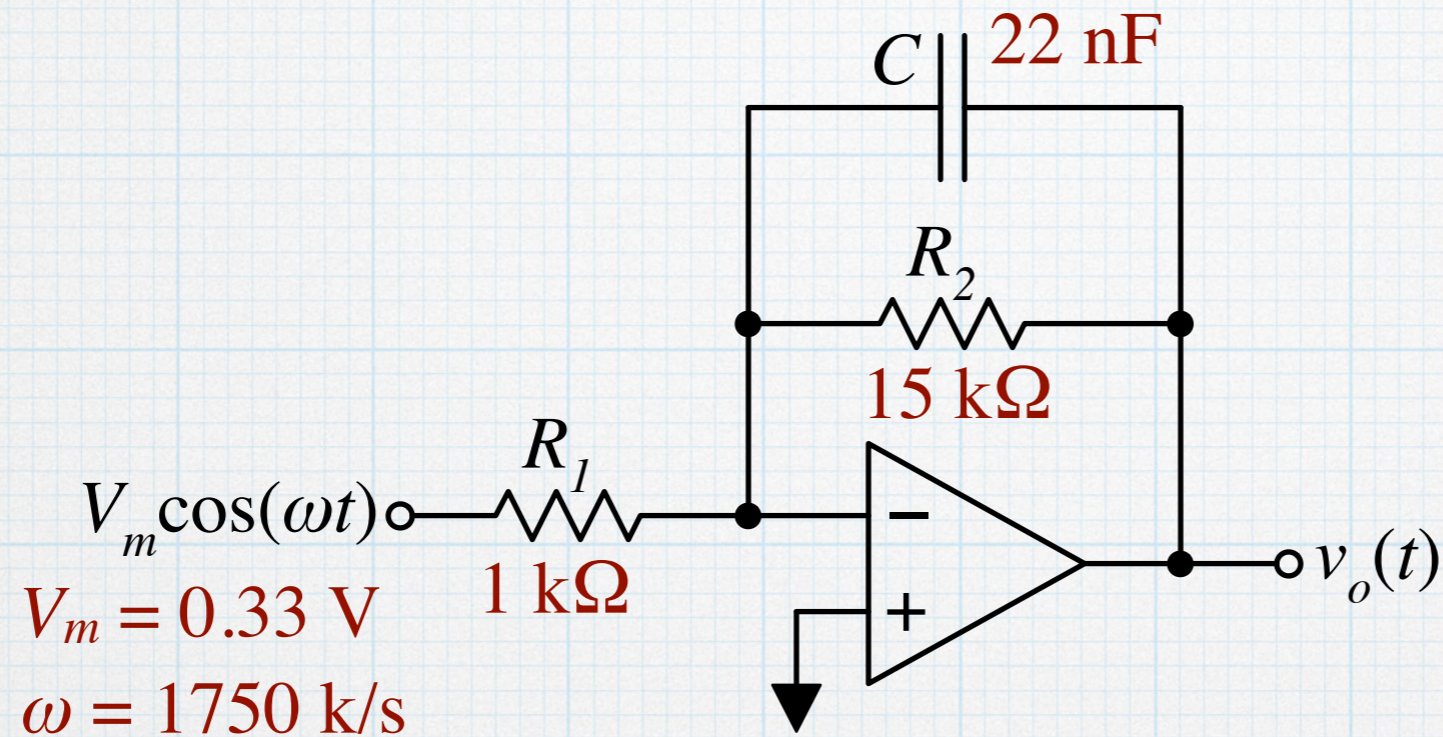
$$= \frac{10 \text{ V} + (10 \text{ V}) e^{j0^\circ}}{1 + (0.667) e^{j90^\circ}} = \frac{20 \text{ V}}{1 + j0.667} = \frac{20 \text{ V}}{1.20 e^{j33.7^\circ}}$$

$$v_a = (16.6 \text{ V}) e^{-j33.7^\circ}$$

Identical to the result obtained in Example 6 using a source transformation.

## Example 9 - node voltage.

Find the complex value of the output voltage in the op-amp circuit shown below. The op-amp is ideal.

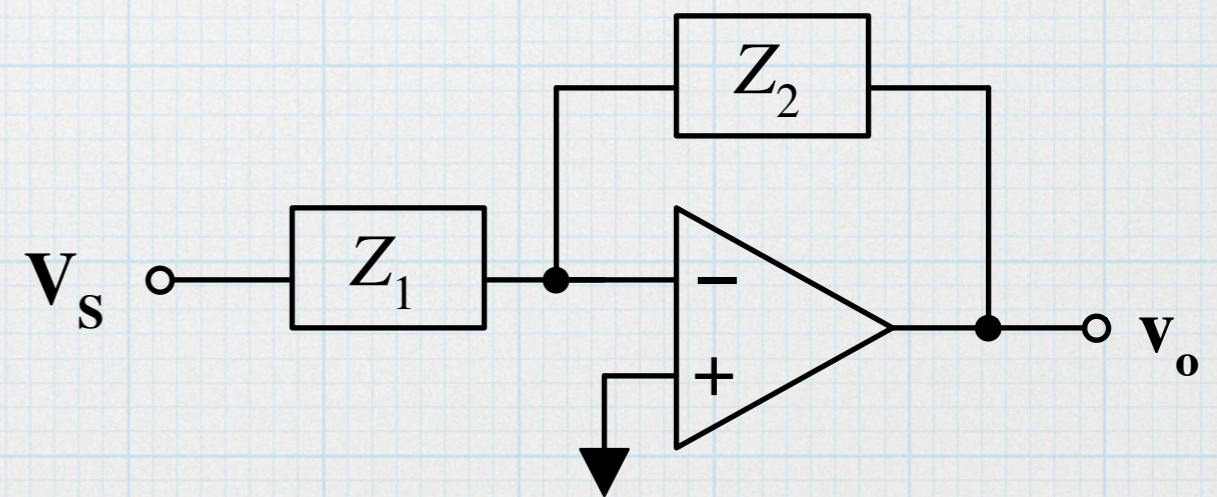


The complex version of the circuit is shown. Using the properties of an ideal op amp with feedback, we can write a node-voltage equation at the inverting terminal:

$$\frac{V_S - v_-}{Z_1} = i_- + \frac{v_- - v_o}{Z_2}$$

$$i_- = 0$$

$$v_- = v_+ = 0$$



$$\frac{\mathbf{V}_S}{Z_1} = -\frac{\mathbf{v}_o}{Z_2}$$

$$\mathbf{v}_o = -\frac{Z_2}{Z_1}\mathbf{V}_S$$

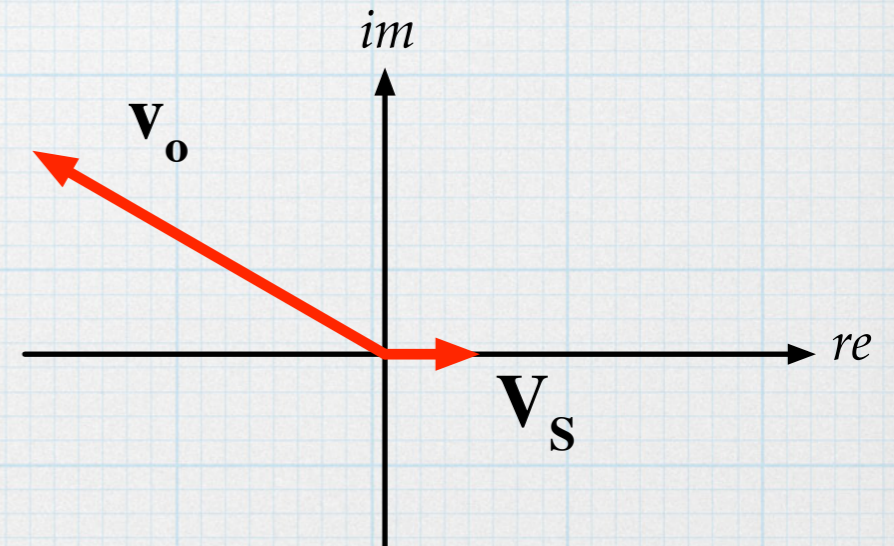
$$Z_2 = R_2 \parallel \left( \frac{1}{j\omega C} \right)$$

$$= \frac{R_2}{1 + j\omega R_2 C} = \frac{15 \text{ k}\Omega}{1 + j0.577}$$

$$Z_1 = R_1 = 1 \text{ k}\Omega$$

$$\mathbf{v}_o = -\frac{15}{1 + j0.577} (0.33 \text{ V}) = \frac{5 \text{ V}}{1.155 e^{j30^\circ}}$$

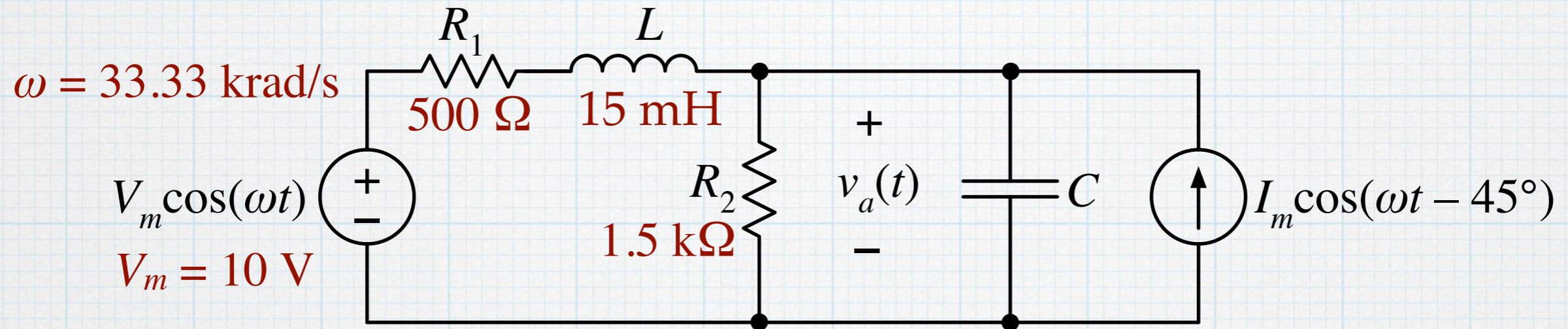
$$\mathbf{v}_o = -(4.33 \text{ V}) e^{-j30^\circ} = (4.33 \text{ V}) e^{j150^\circ}$$



Part of the phase shift is due to the reactance of the capacitor and part is due to the inverting nature of the op-amp.

## Example 11 - mesh current.

Find the complex value of the capacitor voltage in the “two-source, two-impedance” circuit (Example 6) using the mesh-current method.

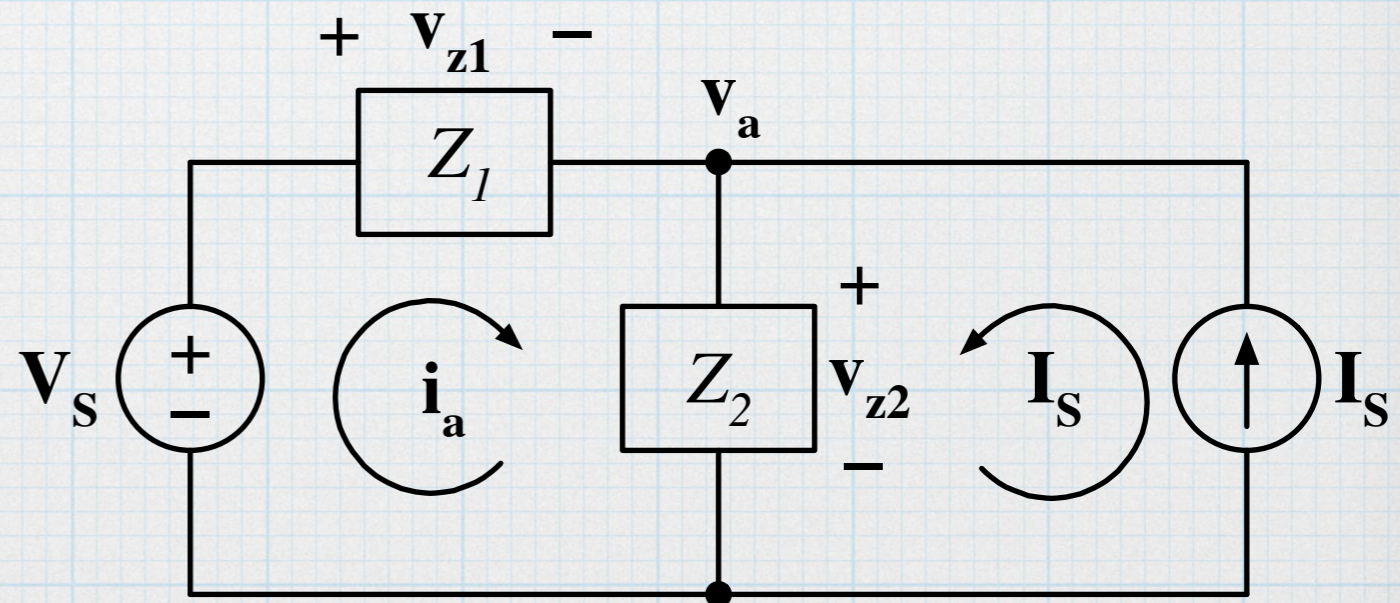


Transform to the complex form of the circuit. The circuit and the values are identical to Example 6. Define the two mesh currents. Of course, the right-hand current is known.

Around the left-hand mesh:

$$V_S - v_{Z1} - v_{Z2} = 0$$

$$V_S - \mathbf{i}_a Z_1 - (\mathbf{I}_S + \mathbf{i}_a) Z_2 = 0$$





$$\mathbf{i}_a = \frac{\mathbf{V}_S - \mathbf{I}_S Z_2}{Z_1 + Z_2}$$

$$= \frac{10 \text{ V} - [(14.14 \text{ mA}) e^{-j45^\circ}] (1060.7 \Omega) e^{-j45^\circ}}{(500 \Omega + j500 \Omega) + (750 \Omega - j750 \Omega)}$$

$$= \frac{10 \text{ V} + j15 \text{ V}}{1250 \Omega - j250 \Omega}$$

Be careful with the sign of the imaginary part.

$$= \frac{(18.03 \text{ V}) e^{j56.3^\circ}}{(1274.8 \Omega) e^{-j11.3^\circ}}$$

$$= (14.14 \text{ mA}) e^{j67.6^\circ}$$

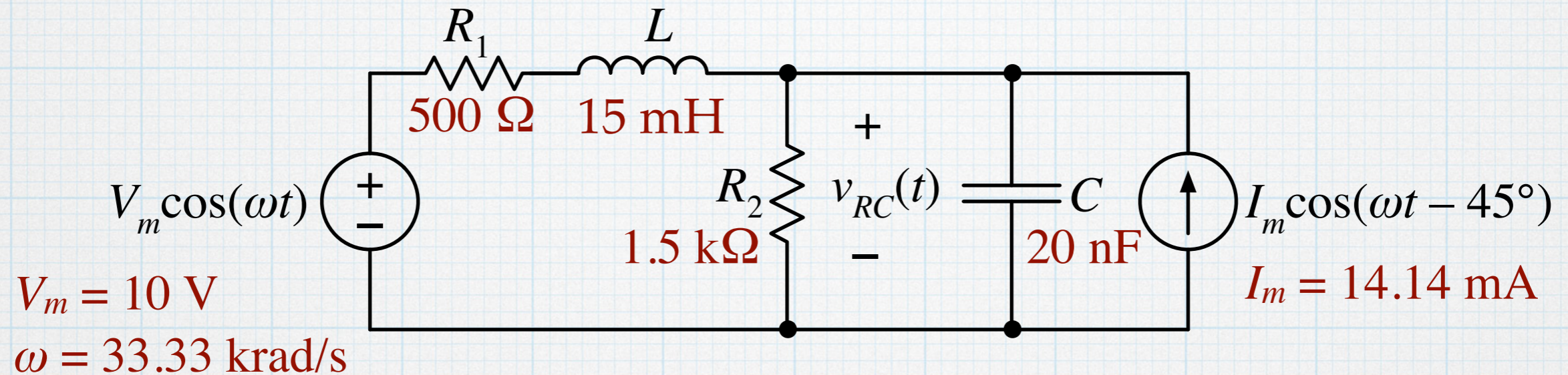
$$\mathbf{v}_a = (\mathbf{i}_a + \mathbf{I}_S) Z_2$$

$$= [(14.14 \text{ mA}) e^{j67.6^\circ} + (14.14 \text{ mA}) e^{-j45^\circ}] [(1060.7 \Omega) e^{-j45^\circ}]$$

$$= (16.6 \text{ V}) e^{-j33.7^\circ} \quad \text{Same.}$$

## Example 13 - superposition.

Find the complex value of the capacitor voltage in the “two-source, two-impedance” circuit using the superposition method.



$$\mathbf{v}'_a = \frac{Z_2}{Z_2 + Z_1} \mathbf{V}_s$$

$$\mathbf{v}'_a = \frac{(1061 \text{ } \Omega) e^{-j45^\circ}}{(1061 \text{ } \Omega) e^{-j45^\circ} + (707 \text{ } \Omega) e^{j45^\circ}} (10 \text{ V})$$

$$Z_1 = (707.1 \text{ } \Omega) e^{j45^\circ}$$

$$Z_2 = (1060.7 \text{ } \Omega) e^{-j45^\circ}$$

$$\mathbf{v}'_a = \frac{Z_2}{Z_2 + Z_1} \mathbf{V}_s$$

$$\mathbf{v}'_a = \frac{(1060.7 \Omega) e^{-j45^\circ}}{(750 \Omega - j750 \Omega) + (500 \Omega + j500 \Omega)} (10 \text{ V})$$

$$= \frac{(1060.7 \Omega) e^{-j45^\circ}}{1250 \Omega - j250 \Omega} (10 \text{ V})$$

$$= \frac{(1060.7 \Omega) e^{-j45^\circ}}{(1274.8 \Omega) e^{-j11.3^\circ}} (10 \text{ V})$$

$$= (8.32 \text{ V}) e^{-j33.7^\circ}$$

$$\mathbf{v}_a'' = \left( Z_1 \parallel Z_2 \right) (\mathbf{I}_S)$$

$$= \left[ (588.3 \Omega) e^{+j11.3^\circ} \right] \left[ (14.14 \text{ mA}) e^{-j45^\circ} \right]$$

$$= (8.32 \text{ V}) e^{-j33.7^\circ}$$

The parallel combination was calculated in the source-transformation version (Example 6) of this problem. We re-use the result here.

$$\mathbf{v}_a = \mathbf{v}_a' + \mathbf{v}_a''$$

$$= (8.32 \text{ V}) e^{-j33.7^\circ} + (8.32 \text{ V}) e^{-j33.7^\circ}$$

$$= (16.6 \text{ V}) e^{-j33.7^\circ} \quad \text{Same.}$$